LECTURE 18:

SHANNON'S THEORM PROOF & PRODUCT CODES

November 13, 2006

• Here's the process (equivalent to generating a random code):

1. We randomly pick *one* codeword, x, which the sender transmits.

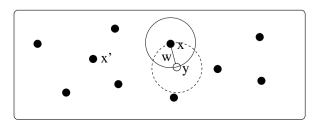
PROVING THE THEOREM WITH RANDOM CODES

- 2. The channel randomly generates an error pattern, n, that is added to x to give the received data, y. Let the number of transmission errors (ie, ones in n) be w.
- 3. We now randomly pick the other M-1 codewords. If the Hamming distance from ${\bf y}$ of all these codewords is greater than w, nearest-neighbor decoding will make the correct choice.
- ullet If the probability that this process leads to a decoding error is $<\epsilon$, then there must be some specific code with error probability $<\epsilon$.
- ullet For large N, the fraction of errors w/N in the transmission is very close to the crossover probability f of the channel:

$$P(f - \beta < w/N < f + \beta) \ge 1 - \epsilon/2$$

SHANNON'S NOISY CODING THEOREM FOR THE BSC

- ullet A BSC with error probability f<1/2 has capacity $1-H_2(f)$.
- Theorem: For any $\eta>0$ and $\epsilon>0$, there is a code (of some length N) whose rate, R, is at least $C-\eta$, and for which the probability that nearest neighbor decoding will fail is less than ϵ .
- Last class we started to give a proof of this, which shows that a randomly chosen code performs quite well and hence that there must be specific codes which also perform quite well (although the proof does not construct such a code).



DECODING WITH TYPICAL ERROR PATTERNS

• The probability that the codeword nearest to y is the correct decoding will be at least as great as the probability that the following sub-optimal decoder decodes correctly:

If there is exactly one codeword \mathbf{x}^* for which $\mathbf{n} = \mathbf{y} - \mathbf{x}^*$ has a "typical" number of ones, then decode to \mathbf{x}^* , otherwise declare that decoding has failed.

- This sub-optimal decoder can fail in two ways:
 - The correct decoding, x, may correspond to an error pattern, n = y x, that is not "typical".
 - Some other codeword, \mathbf{x}' , may exist for which the error pattern $\mathbf{n}'=\mathbf{y}-\mathbf{x}'$ is typical.
- ullet The total probability of decoding failure is less than the sum of the probabilities of failing in these two ways. We will try to limit each of these to $\epsilon/2$.

ullet We can choose N large enough to ensure the actual error pattern will be non-typical with probability less than $\epsilon/2$:

$$P(f - \beta < w/N < f + \beta) \ge 1 - \epsilon/2$$

- \bullet We now need to limit the probability that some other codeword also corresponds to one of the J typical error patterns. J is bounded by $I < 2^{N(H_2(f) + \beta \log_2((1-f)/f))}$
- \bullet For a random codeword \mathbf{x}' (other than the one actually transmitted), the difference $(\mathbf{x}'-\mathbf{y})$ will contain 0s and 1s that are independent and equally likely. It will be "typical" with probability $J/2^N \ < \ 2^{-N(1-H_2(f)-\beta\log_2((1-f)/f))}$
- ullet The probability that any of the M-1 incorrect codewords will differ from ${f y}$ by a typical error pattern is bounded by M times this. We need this bound to be less than $\epsilon/2$, ie

$$M 2^{-N(1-H_2(f)-\beta \log_2((1-f)/f))} < \epsilon/2$$

- Recall that for a code to be guaranteed to correct up to t errors, it's minimum distance must be at least 2t + 1.
- What's the minimum distance for the random codes used to prove the noisy coding theorem?
- ullet A random N-bit code is very likely to have minimum distance $d \leq N/2$ if we pick two codewords randomly, about half their bits will differ. So these codes are likely *not guaranteed* to correct patterns of N/4 or more errors.
- ullet A BSC with error probability f will produce about Nf errors. So for f>1/4, we expect to get more errors than the code is guaranteed to correct. Yet we know these codes are good!
- **Conclusion:** A code may be able to correct *almost all* patterns of *t* errors even if it can't correct *all* such patterns.

FINISHING THE PROOF

į

- ullet Finally, we need to pick eta, M, and N so that the two types of error have probabilities less than $\epsilon/2$, and the rate, R is at least $C-\eta$.
- We will let $M = 2^{\lceil (C \eta)N \rceil}$, and make sure N is large enough that $R = \lceil (C \eta)N \rceil/N < C$ [and $P(f \beta < w/N < f + \beta) \ge 1 \epsilon/2$].
- $\begin{array}{c} \bullet \text{ With this value of } M \text{, we need} \\ 2^{\lceil (C-\eta)N \rceil} \ 2^{-N(1-H_2(f)-\beta \log_2((1-f)/f))} < \epsilon/2 \\ \Rightarrow 2^{-N(1-H_2(f)-\lceil (C-\eta)N \rceil/N-\beta \log_2((1-f)/f))} < \epsilon/2 \end{array}$
- The channel capacity is $C=1-H_2(f)$, so that $1-H_2(f)-\lceil (C-\eta)N\rceil/N=C-R$ is positive.
- For a sufficiently small value of β (which we can get with large N), $1-H_2(f)-\lceil (C-\eta)N\rceil/N-\beta\log_2((1-f)/f)$ will also be positive. With this β and a large enough N, the probabilities of both types of error will be less than $\epsilon/2$, so the total error probability will be less than ϵ .

PRODUCT CODES

7

- A product code is formed from two other codes C_1 , of length N_1 , and C_2 , of length N_2 . The product code has length N_1N_2 .
- ullet We can visualize the N_1N_2 symbols of the product code as a 2D array with N_1 columns and N_2 rows.
- Definition of a product code:
 An array is a codeword of the product code if and only if
- all its rows are codewords of \mathcal{C}_1
- —all its columns are codewords of $\mathcal{C}_2^{\ \ _{N_2-K_2}}$

K₁ N₁ - K₁

Bits of the message being encoded Check bits computed from the rows

Check bits computed from the columns Check bits computed from the check bits

- ullet We will assume here that \mathcal{C}_1 and \mathcal{C}_2 are linear codes, in which case the product code is also linear. (Can you see why?)
- The product codeword (as a matrix) is the outer product of the two original codewords (as vectors).

10

- Suppose C_1 is an $[N_1, K_1]$ code and C_2 is an $[N_2, K_2]$ code. Then their product will be an $[N_1N_2, K_1K_2]$ code.
- Suppose C_1 and C_2 are in systematic form. Here's a picture of a codeword of the product code:

	κ_1	$N_1 - K_1$
К ₂	Bits of the message being encoded	Check bits computed from the rows
N ₂ - K ₂	Check bits computed from the columns	Check bits computed from the check bits

- The dimensionality of the product code is not more than K_1K_2 , since the message bits in the upper-left determine the check bits.
- We'll see that the dimensionality equals K_1K_2 by showing how to find correct check bits for any message.

- Products of even small codes have lots of check bits, so decoding directly may be infeasible.
- But if \mathcal{C}_1 and \mathcal{C}_2 can easily be decoded, we can decode the product code by first decoding the rows (replacing them with the decoding), then decoding the columns. (Or the other way around.)
- This will usually **not** be a nearest-neighbor decoder (and hence will be sub-optimal, assuming a BSC and equally-likely messages).

ENCODING PRODUCT CODES

ĵ

- Here's a procedure for encoding messages with a product code:
- 1. Put K_1K_2 message bits into the upper-left K_2 by K_1 corner of the N_2 by N_1 array (using the outer product of the messages).
- 2. Compute the check bits for the first K_2 rows, according to C_1 .
- 3. Compute the check bits for all N_1 columns, according to C_2 .
- After this, all the columns will be codewords of C_2 , since they were given the right check bits in step (3). The first K_2 rows will be codewords of C_1 , since they were given the right check bits in step (2). But are the last $N_2 K_2$ rows codewords of C_1 ?
- ullet Yes! Check bits are linear combinations of message bits. So the last N_2-K_2 rows are linear combinations of earlier rows. Since these rows are in \mathcal{C}_1 , their combinations are too.

MINIMUM DISTANCE OF PRODUCT CODES

1:

• If C_1 has minimum distance d_1 and C_2 has minimum distance d_2 , then the minimum distance of their product is d_1d_2 .

• Proof:

Let \mathbf{u}_1 be a codeword of \mathcal{C}_1 of weight d_1 and \mathbf{u}_2 be a codeword of \mathcal{C}_2 of weight d_2 . Build a codeword of the product code by putting \mathbf{u}_1 in row i of the array if \mathbf{u}_2 has a 1 in position i. Put zeros elsewhere. This codeword has weight d_1d_2 .

The new codeword is the outer product of the vectors \mathbf{u}_1 and \mathbf{u}_2 .

u1	u2	0	1	0	0	0	1	0	0	0	
							1			:	:
0			1.							ļ	:
				ļ						ļ	ŀ
0				ļ			ŀ.:			}	ŀ
0			l.	 			Н.				
0											
0				 						}	ŀ
1			7	ļ			1			}	ŀ
0			•				•	-		· · ·	
0							•			ļ	١

ullet Furthermore, any non-zero codeword must have at least this weight. It must have at least d_2 rows that aren't all zero, and each such row must have at least d_1 ones in it.

- Let \mathcal{C} be an [N, K] code of minimum distance d (guaranteed to correct $t = \lfloor (d-1)/2 \rfloor$ errors).
- \bullet How good is the code obtained by taking the product of ${\mathcal C}$ with itself p times?

Length: $N_p = N^p$

Rate: $R_p = K^p/N^p = (K/N)^p \to 0$

Distance: $d_p = d^p$

Relative distance: $\rho_p = d_p/N_p = (d/N)^p \to 0$

- ullet The code can correct up to about $d_p/2$ errors, corresponding to a proportion of errors of $ho_p/2$.
- ullet For a BSC with error probability f, we expect that for large N, the proportion of erroneous bits in a block will be very close to f. (The Law of Large Numbers once again.)

Why use Products of Codes?

13

- ullet The analysis above shows that for large N, these product codes are both *unlikely* to correct all errors, and also that they have a low rate (approaching zero)!
- Futhermore, they are hard to decode in an exact (maximum likelihood) way, so we have to use an approximate decoder.
- So why would we ever use them?
- One advantage of product codes: They can correct some *burst errors* errors that come together, rather than independently.